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UTILITY PATENT APPLICATION **TRANSMITTAL**

ATI-000150BT Attorney Docket No.

First Inventor or Application Identifier Feliks Dujmenovic

Title IMAGE REJECTION MIXING IN WIDEBAND APPLICATIONS

(Only for new nonprovisional applications under 37 C.F.R. § 1.53(b)) Express Mail Label No. | EL703379723US

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Application Number	Not Yet Known					
Filing Date	Not Yet Known					
First Named Inventor	Feliks Dujmenovic					
Examiner Name	Not Yet Known					
Group / Art Unit	Not Yet Known					
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1. BASIC FILING FEE Large Entity Small Entity	117 870 217 435 Extension for reply within third month						
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IMAGE REJECTION MIXING IN WIDEBAND APPLICATIONS

BACKGROUND

This invention generally relates to wideband communication systems. In particular, the invention relates to image suppression in received wideband communications.

Wideband communication systems, such as television and radio communication systems, use a wide frequency spectrum to communicate information. Typically, the wide spectrum is divided into a group of assigned radio frequencies for carrying the information.

In a heterodyne receiver, the received signal is amplified by a radio frequency (RF) amplifier and mixed with an adjustable local oscillator (LO) signal to produce intermediate frequency (IF) signal. One problem with mixer circuits is the generation of image frequency signals.

When two signals are mixed, signal components are produced at the sum and difference of the two signals, and their harmonics. Equation 1 illustrates the potential signal components where F_{LO} is a local oscillator frequency being mixed with a radio frequency, F_{RF} and m and n are integers or zero.

 $\pm m$ Frf $\pm n$ Flo

Equation 1

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Since input circuits typically have limited selectivity in the mixing process, undesired interference may create harmful products, as illustrated by Equation 2. F_I is an undesired interference frequency and m, n and p are an integer or zero.

 $\pm \ m \ F_{\text{RF}} \pm n \ F_{\text{LO}} \pm p \ F_{\text{I}}$

Equation 2

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For example, when an intermediate frequency signal, F_{RF} , is mixed with a local oscillator, F_{LO} , the result will produce signals at two frequencies, $F_{LO} + F_{RF}$ and $F_{LO} - F_{RF}$. One of the two signals is the desired frequency and the other is at an undesired frequency. Additionally, based on the quality of the mixers, undesirable harmonics, such as $F_{LO} + 2F_{RF}$, could also be produced. In a wideband communication system utilizing multiple frequencies, the frequencies that these undesired signals fall upon may be the same frequency used for another information signal. As a result, the undesired signal produces an image on the desired information signal at that frequency.

One approach to removing the image signal from the desired signal is by filtering. Typically, filtering only reduces the magnitude of the image signal by less than 30 decibels (dB). In some applications, a 30 dB attenuation is not sufficient.

Another approach is to use an image rejection mixer 10, as illustrated in Figure 1. A received radio frequency signal is input into the image rejection mixer 10. The received signal is input to an in-phase mixer 14 and a quadrature phase mixer 12. A local oscillator (LO) 11 generates a carrier signal. The carrier signal is input into the quadrature phase mixer 12 to produce a demodulated quadrature phase signal and into a 90 degree phase shift device 16, such as a RC–CR circuit, to produce an in-phase carrier. The in-phase carrier is input into the in-phase mixer 12 to produce a demodulated in-phase signal. The demodulated quadrature phase signal is subsequently delayed by a 90 degree phase shift device 22. An adder 20 combines the phase delayed quadrature phase signal to the in-phase signal to produce the desired signal. Typically, the image signal will be out of phase with the desired signal. As a result, the combining cancels the image signal leaving only the desired signal.

To illustrate, W_{RF} is the frequency of the received signal and W_{LO} is the frequency at the local oscillator. The image rejection mixer 10 processes the desired signal as follows. I(t) is the demodulated in-phase signal. Q(t) is the demodulated quadrature phase signal. Q'(t) is the phase delayed quadrature phase signal. O(t) is the subtracted signal.

$$\begin{split} I(t) &= \cos{(W_{RF}t)} * \cos{(W_{LO}t)} = \\ &= \frac{1}{2} [\cos{(W_{RF} - W_{LO})} t + \cos{(W_{RF} + W_{LO})} t \] \ , \ where \ W_{RF} - W_{LO} > 0 \end{split}$$
 Equation 3

The cos (WRF + WLO) t is removed by low-pass filtering.

$$Q(t) = \cos (W_{RF}t) * \cos (W_{LO} t - \pi/2) =$$

$$= \frac{1}{2} \cos [(W_{RF} - W_{LO})t + \pi/2] = -\frac{1}{2} \sin (W_{RF} - W_{LO})t$$

Equation 4

$$Q'(t) = -\frac{1}{2} \sin [(W_{RF} - W_{LO})t - \pi/2] = \frac{1}{2} \cos (W_{RF} - W_{LO})t$$

Equation 5

$$O(t) = I(t) + O'(t) = \cos(W_{RF} - W_{LO})t$$

Equation 6

The image rejection mixer 10 processes the image signal which is inverted with respect to the desired signal as follows.

$$I(t) = \frac{1}{2} \cos \left[-(W_{RF} - W_{LO}) t \right] = \frac{1}{2} \cos (W_{RF} - W_{LO}) t$$

Equation 7

$$Q(t) = \frac{1}{2} \cos \left[-(W_{RF} - W_{LO})t + \pi/2 \right] = \frac{1}{2} \sin (W_{RF} - W_{LO})t$$

Equation 8

$$Q'(t) = \frac{1}{2} \sin [(W_{RF} - W_{LO})t - \pi/2] = -\frac{1}{2} \cos (W_{RF} - W_{LO})t$$

Equation 9

$$O(t) = I(t) + Q'(t) = \frac{1}{2}\cos (W_{RF} - W_{LO})t - \frac{1}{2}\cos (W_{RF} - W_{LO})t = 0$$

Equation 10

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Accordingly, after image cancellation, the desired signal is recovered, Equation 6, and the image signal is canceled, Equation 10.

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Due to variations in the tolerances of the resistors and capacitors in the phase shift device 16 and temperature variations, the phase differential between the carrier and the quadrature carrier may not be maintained at an ideal 90 degrees, degrading performance of the image mixer.

Accordingly, it is desirable to have alternate approaches for image suppression in received wideband signals.

SUMMARY

A ring oscillator produces an in-phase and quadrature phase radio frequency signal. A first mixer mixes the in-phase signal with a received signal. A second mixer mixes the quadrature phase signal with the received signal. A combiner, operatively coupled to the first and second mixers, produces an image cancelled signal.

BRIEF DESCRIPTION OF THE DRAWINGS

Figure 1 is an image rejection mixer.

Figure 2 is an image rejection mixer using a ring oscillator.

Figure 3 is a delay cell.

Figure 4 is a RC-CR circuit.

BRIEF DESCRIPTION OF THE PREFERRED EMBODIMENT(S)

Figure 2 illustrates an image rejection mixer 24 for use in a wideband communication system receiver, such as a television tuner or radio tuner. Although the image rejection mixer 24 is shown using balanced mixers 28, 30, such as a Gilbert cell, other mixers may be used. Balanced mixers are desirable due to their noise suppression quality. A received RF signal is buffered by a buffer 58 so that the RF signal as outputted by the buffer 58 is shifted to a desired level for the in-phase and quadrature mixers 30, 28. The buffered RF signal is inputted into both the in-phase and quadrature phase mixers 30, 28. A ring oscillator 32 produces an in-phase carrier which is input into the in-phase mixer 30 and a quadrature phase carrier which is input to the quadrature

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phase mixer 28. The demodulated quadrature phase signal is subsequently delayed by a 90° phase delay device 38, such as an RC–CR circuit. The delayed quadrature phase signal is combined using an adder 40 to the in-phase signal to produce the desired signal with the image signal cancelled.

By using the ring oscillator 32, the in-phase and quadrature carrier are maintained at a near ideal 90 degree phase difference. Since the demodulated signals are at low frequencies, a simpler 90 degree phase shift device 38, such as an RC-CR circuit, may be used without degrading the image rejection mixer's performance. Using the ring oscillator 32, a 60 dB attenuation in the image signal is achieved.

One ring oscillator 32 for producing the in-phase and quadrature phase carrier uses four delay cells 44₁-44₄, as shown in Figure 2. Each delay cell 44₁-44₄ delays the input signal by 45 degrees. As a result, the first delay cell 44₁ produces a 45 degree phase delay. The second delay cell delays the 45 degree delayed signal by another 45 degrees, totaling 90 degrees. The output of the second delay cell 44₂ is inverted as to shift the phase by 180 degrees prior to being input into the third delay cell 44₃. The third delay cell 44₃ delays the 270 degree delayed signal by 45 degrees, totaling 315 degrees. The fourth cell 44₄ delays the output of the third cell 44₃ by 45 degrees totaling 360 degrees creating the oscillation. Although the output of the second cell 44₂ is shown as used for the in-phase carrier and the output of the fourth cell 44₄ for the quadrature phase carrier, any of the outputs or inputs separated by 90 degrees may be used.

One circuit for use as a delay cell 44₁-44₄ is shown in Figure 3 using CMOS circuitry. Six n-type MOSFETS 46₁-46₆ and two p-type MOSFETS 48₁, 48₂ are configured as shown in Figure 3. The inputs to the gates of the two bottom n type MOSFETS 46₁, 46₂ are connected in series to form the bias, control frequency, to the delay cells 44₁-44₄. The gates of the two lower n type MOSFETS 46₃, 46₆ form the input of the delay cells 44₁-44_n. The drains of those MOSFETS 46₃, 46₆, form the output. This CMOS circuit produces an output having a phase delayed by 45 degrees from the input.

A simple 90 degree phase shift device 38 may be used for the quadrature phase demodulated signal, such as a RC-CR circuit as shown in Figure 4. The demodulated

quadrature phase signal is input into the delay device 38. The input of the delay device 38 is coupled to a series connected resistor 50 and capacitor 52 connected in parallel to a series connected capacitor 54 and resistor 56. An output of 90 degrees of phase delay is produced at the nodes between the series connected resistors 50, 56 and capacitors 52, 54.

* *

CLAIMS

What is claimed is:

- 1. An apparatus for canceling an image signal from a received radio frequency signal, the apparatus comprising:
- a ring oscillator for producing a radio frequency signal having in-phase and quadrature phase components;
- a first mixer having inputs configured to receive the in-phase component and the received radio frequency signal and outputting an in-phase signal;
- a second mixer having inputs configured to receive the quadrature phase component and the received radio frequency signal and outputting a quadrature phase signal;
- a phase shift device coupled with one of the mixers for receiving an output of the one mixer and outputting a phase shifted signal; and
- a combiner, operatively coupled to the other of the mixers and said phase shift device, for producing an image cancelled signal.
- 2. The apparatus of claim 1 wherein the phase shift device is coupled to the second mixer.
- 3. The apparatus of claim 2 wherein the phase shift device shifts a phase of the second mixer output by ninety degrees.
- 4. The apparatus of claim 1 wherein the ring oscillator comprises four delay cells, an output of each delay cell is coupled to an output of another of the delay cells.
- 5. The apparatus of claim 4 wherein each delay cell delays its input by forty-five degrees and one of the couplings is cross-coupled so that the output of one of the delay cells is inverted prior to input into another of the delay cells.

- 6. The apparatus of claim 1 wherein the first mixer and the second mixer are gilbert cells.
- 7. A receiver for use in a wideband communication system, the receiver capable of canceling an image signal from a received radio frequency signal, the receiver comprising:
- a ring oscillator for producing a radio frequency signal having in-phase and quadrature phase components;

first mixing means for mixing the in-phase component with the received radio frequency signal and outputting an in-phase signal;

second mixing means for mixing the quadrature phase component with the received radio frequency signal and outputting a quadrature phase signal;

means for receiving one of the mixer's phase signals and outputting a phase shifted signal; and

means for combining the phase shifted signal with the phase signal other than the one phase signal to produce an image canceled signal.

- 8. The receiver of claim 7 wherein the means for outputting a phase shifted signal shifts the one phase signal by ninety degrees in phase.
- 9. The receiver of claim 7 wherein the ring oscillator comprises four delay cells, an input of each delay cell is coupled to an output of another of the delay cells.
- 10. The receiver of claim 9 wherein each delay cell delays its input by forty-five degrees and one of the couplings is cross-coupled so that the output of the delay cell is inverted prior to input into another of the delay cells.
- 11. The receiver of claim 7 wherein the means for outputting a phase shift signal is coupled to the first mixing means.

- 12. The receiver of claim 7 wherein the first and second mixing means comprises a gilbert cell.
- 13. A method for canceling an image signal from a received radio frequency signal, the method comprising:

providing a ring oscillator;

producing a radio frequency signal having in-phase and quadrature phase components with the ring oscillator;

mixing the in-phase component and the received radio frequency signal to produce an in-phase signal;

mixing the quadrature phase component and the received radio frequency signal to produce a quadrature phase signal;

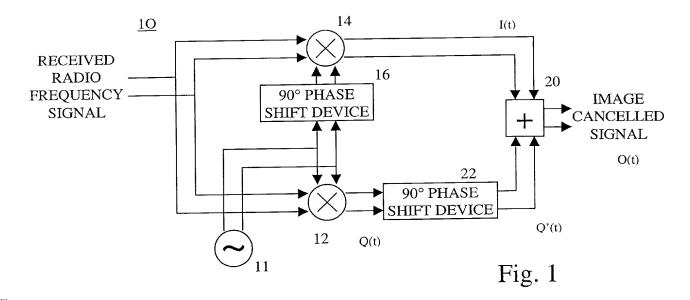
shifting a phase of one of the phase signals produced by mixing the components to produce a phase shifted signal; and

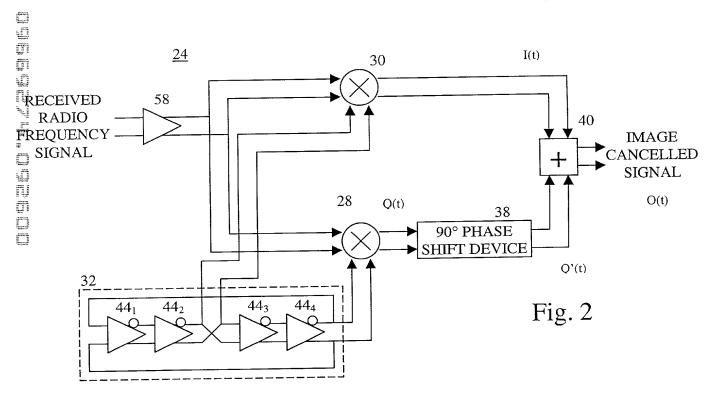
combining the phase shifted signal with the phase signal produced by mixing the components other than the one phase signal to produce an image canceled signal.

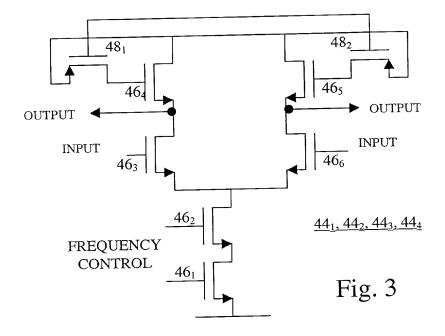
- 14. The method of claim 13 wherein the one phase signal is the quadrature phase signal.
- 15. The method of claim 13 wherein the phase shifting is by ninety degrees in phase.

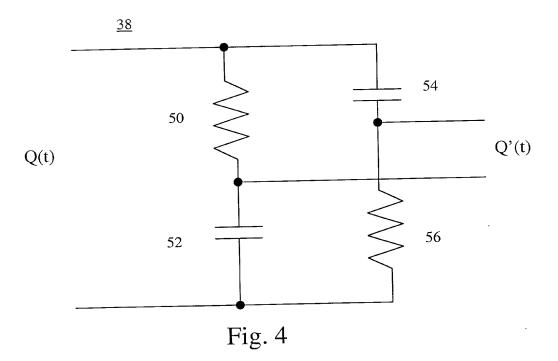
ABSTRACT

A ring oscillator produces an in-phase and quadrature phase radio frequency signal. A first mixer mixes the in-phase signal with a received signal. A second mixer mixes the quadrature phase signal with the received signal. A combiner, operatively coupled to the first and second mixers, produces an image cancelled signal.









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ATI-000150BT Attorney Docket Number **DECLARATION FOR UTILITY OR** Feliks Dujmenovic First Named Inventor DESIGN COMPLETE IF KNOWN PATENT APPLICATION Application Number (37 CFR 1.63) Not Yet Known Not Yet Known Filing Date ☑ Declaration ☐ Declaration Not Yet Known OR Submitted after Initial Submitted **Group Art Unit**

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